COMPRESSIVE SENSING IN NONSTATIONARY ARRAY PROCESSING
USING BILINEAR TRANSFORMS

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ABSTRACT

Compressive sensing (CS) has successfully been applied to reconstruct sparse signals and images from few observations. For multi-component nonstationary signals characterized by instantaneous frequency laws, the sparsity exhibits itself in the time-frequency domain as well as the ambiguity domain. In this paper, we examine CS in the context of nonstationary array processing. We show that the spatial averaging of the ambiguity function across the array improves the CS performance by reducing both noise and cross-terms. The corresponding time-frequency distribution which is reconstructed through $L_1$ minimizations yields significant improvement in time-frequency signature localizations and characterizations.

1. INTRODUCTION

Compressive sensing (CS) has successfully been applied to electromagnetic (EM) sensing problems using multiple antennas, including synthetic aperture radar (SAR) and indoor imaging [1, 2, 3, 4, 5]. In this paper, we apply CS to sparse signal time-frequency representation using antenna arrays. We consider nonstationary signals with instantaneous frequency (IF) characterization. These signals are encountered in many applications including radar, sonar, audio, and biomedicine. The spatial degrees-of-freedom, offered by the multi-sensor system configuration, are used to reduce noise and cross-terms in the ambiguity domain without altering the signal sparsity profile in the joint-variable bilinear representation. This leads to higher time-frequency signature resolution when applying sparse signal reconstruction techniques.

In bilinear time-frequency analyses of nonstationary signals, the existence of cross-terms is a major problem that compromises, and even obscures, the proper identification of signal energy distribution in the time-frequency domain. To mitigate the effect of cross-terms, it is desirable to design a mask that only captures the auto-term components in the ambiguity domain. This concept underlines interference-reducing kernels in the model-free Cohen’s class of time-frequency distributions [6]. While many time-frequency kernels, such as the well-known Choi-Williams distribution [7], are data-independent, others, such as the adaptive optimal kernel, optimize the kernel based on the signal characteristics [8]. In either case, cross-term mitigation is achieved at the expense of reduced resolution in the time-frequency domain. In multi-sensor systems, on the other hand, it is shown in [9, 10] that the averaging of the time-frequency distributions, each is computed at a single sensor, amounts to spatial filtering of the cross-terms. The level of cross-term reduction depends on the spatial correlation between the corresponding source signals. Mathematically, averaging the time-frequency distributions across the array amounts to taking the trace of the spatial time-frequency distribution (STFD) matrix [11, 12].

In this paper, we examine quadratic time-frequency analysis of nonstationary signals in a CS multi-sensor array platform. The focus is on proper identifications of auto-term regions in the ambiguity domain which is key to accurate sparse signal reconstruction. We first show that, because the ambiguity function and the time-frequency distribution are related by the two-dimensional (2-D) Fourier transform, spatial averaging of the ambiguity functions correspond to the same averaging operation in the time-frequency domain, both lead to cross-term and noise reductions. This sensor averaging operation, combined with lowpass filtering in ambiguity domain, which emphasizes the signal auto-terms, enhances the signal-to-interference-plus-noise ratio (SINR) without altering the sparsity profile of the signal joint-variable representation. Compressive sensing performed on the weighted/ masked ambiguity function observations yields improved time-frequency signal energy localizations. This improvement is witnessed over single-sensor based approaches as well as when compared to performing $L_2$ norm minimization and 2-D Fourier transform.

The following notations are used in this paper. A lower (upper) case bold letter denotes a vector (matrix). $(\cdot)^*$, $(\cdot)^T$ and $(\cdot)^H$ respectively denote complex conjugation, transpose, and conjugate transpose (Hermitian) operations. $I_n$ expresses the $n \times n$ identity matrix. $\mathcal{F}(\cdot)$ and $\mathcal{F}^{-1}(\cdot)$ respectively represent the 2-D Fourier transform and 2-D inverse Fourier transform. $\| \cdot \|_1$ and $\| \cdot \|_2$ respectively denote the $L_1$ and $L_2$ norm operations. $\dot{\phi}(t) = d\phi(t)/dt$ represents the derivative of $\phi(t)$ with respect to time. In addition, $\mathbb{C}^{N \times M}$ denotes the complete set of $N \times M$ complex entries.

2. SIGNAL MODEL

Consider $K$ narrowband nonstationary signals impinging on an array consisting of $N$ sensors. The $N \times 1$ received data vector $y(t)$ and the $K \times 1$ source signal vector $d(t)$ are related by

$$y(t) = Hd(t) + n(t), \quad (1)$$

where $t$ is the time index, $H = [h_1, h_2, ..., h_K] \in \mathbb{C}^{N \times K}$ is the mixing matrix that holds the steering vectors of the $K$
signals, and \(h_y\) is the spatial signature for the \(q\)th source, \(d_q(t)\), that may arrive with or without an angular bearing. \(T\) is the number of observation samples. Each element of \(d(t) = [d_1(t), d_2(t), \ldots, d_K(t)]^T\) is assumed to be a mono-component signal. \(n(t) \in \mathbb{C}^{N \times 1}\) is an additive noise vector that consists of independent and identically distributed (i.i.d.) zero-mean, white and complex Gaussian distributed processes with variance \(\sigma_n^2\). The noise elements are assumed to be independent of the signals, which are assumed to be deterministic.

3. TIME-FREQUENCY DISTRIBUTION AND AMBIGUITY FUNCTION

For a non-stationary signal \(x(t)\), its ambiguity function is defined as

\[
A_{xx}(\theta, \tau) = \int_{-\infty}^{\infty} x \left( u + \frac{T}{2} \right) x^* \left( u - \frac{T}{2} \right) e^{-j\theta u} du,
\]

where \(\theta\) and \(\tau\) are the frequency lag and the time lag, respectively. Similarly, the time-frequency distribution, within the Cohen’s class, is defined as the 2-D Fourier transform of the ambiguity function [6]

\[
D_{xx}(t, f) = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} \phi(\theta, \tau) A_{xx}(\theta, \tau) e^{-j2\pi ft - j2\pi\tau\theta} d\theta d\tau,
\]

where \(t\) and \(f\) represent the time and frequency indexes, respectively, and \(\phi(\theta, \tau)\) is the time-frequency kernel. A time-frequency kernel is often designed to emphasize auto-terms and mitigate cross-terms interference. A kernel is typically a 2-D low-pass filter in the ambiguity domain. The reasoning behind this filter is that the auto-terms are generally concentrated around the origin, whereas the cross-terms are positioned away from the origin. However, the exact auto- and cross-term distributions depend on the waveforms and thus may differ significantly for different signals.

For two non-stationary signals \(x_i(t)\) and \(x_k(t)\), their cross-term ambiguity function can be defined as

\[
A_{x_i x_k}(\theta, \tau) = \int_{-\infty}^{\infty} x_i \left( u + \frac{T}{2} \right) x_k^* \left( u - \frac{T}{2} \right) e^{-j\theta u} du,
\]

and their cross-term time-frequency distribution is defined as

\[
D_{x_i x_k}(t, f) = \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} \phi(\theta, \tau) A_{x_i x_k}(\theta, \tau) e^{-j2\pi ft - j2\pi\tau\theta} d\theta d\tau.
\]

An element of vector \(y(t)\) in (1) is a mixture of multiple signal arrivals and therefore becomes a multi-component signal. Therefore, its ambiguity function and time-frequency distribution, due to their bilinear nature, contain auto-terms and cross-terms. The cross-terms often obscure the identification of the auto-term regions, which reveal the true signal power distribution and, as such, the underlying non-stationary signal characteristics. Cross-term mitigation or reduction is also important for sparse signal reconstruction, as shown in the simulation section.

4. SPATIAL AVERAGING

The ambiguity function and time-frequency distribution are two different representations that map a time-domain non-stationary signal to 2-D joint-variable representations that allow more flexible processing and filtering. The proper selection of signal auto-terms amounts to signal-to-noise ratio (SNR) enhancement and enables source signature discriminations. The latter provides improved direction-of-arrival estimation, source separation, and waveform recovery [13, 14].

For most commonly used time-frequency kernels, the auto-terms time-frequency distributions are real. These terms are also positive for meaningful time-frequency points where the signal energy is concentrated. On the other hand, cross-terms are complex in general, and their values depend on the relative phase between the contributing signals. As such, the simple averaging of time-frequency distributions over different antennas enhances the auto-terms. The same averaging operation reduces cross-terms if the spatial correlation between the contributing signals is low.

With the focus on cross-term suppression, we consider a noise-free scenario, where the \(ith\) diagonal element of \(\mathbf{D}_{yy}(t, f)\) is expressed as [10]

\[
D_{yy}(t, f) = \sum_{l=1}^{K} \sum_{k=1}^{K} a_{il} a_{ik}^* D_{dl, dk}(t, f),
\]

where \(a_{il}\) is the \((i, l)th\) element of \(\mathbf{A}\). Averaging the \(N\) diagonal elements of \(\mathbf{D}_{yy}(t, f)\) thus becomes

\[
\mathbf{D}_{yy}(t, f) = \frac{1}{N} \sum_{i=1}^{N} D_{yi, yi}(t, f) = \sum_{l=1}^{K} \sum_{k=1}^{K} \beta_{l, k} D_{dl, dk}(t, f),
\]

where \(D_{dl, dk}(t, f)\) denotes the cross-term between signal component \(d_l(t)\) and \(d_k(t)\), and

\[
\beta_{l, k} = \frac{1}{N} \sum_{i=1}^{N} a_{il} a_{ik}^*.
\]

is the spatial correlation, defined in the \(N\)-element array, of the return signals from targets \(l\) and \(k\).

The ambiguity function, whether in the form of auto-terms or cross-terms, is in general complex. However, the auto-term of a signal component is invariant with the phase difference across the sensor array due to differences in the relative propagation delays. On the other hand, different signal components originated from different sources will yield different phase delays due to differences in their angular locations. Therefore, spatial averaging of the ambiguity functions bears the same spatial filtering capability for cross-term reduction as the averaging of time-frequency distributions. This can be easily inferred because the 2-D Fourier transform associating the ambiguity function and time-frequency distribution is linear. As a result, the ambiguity function averaged over the \(N\) array sensors can be expressed as

\[
\hat{A}_{yy}(\theta, \tau) = \frac{1}{N} \sum_{i=1}^{N} A_{yi, yi}(\theta, \tau) = \sum_{l=1}^{K} \sum_{k=1}^{K} \beta_{l, k} A_{dl, dk}(\theta, \tau),
\]

(9)
where $A_{d_i d_j}(\theta, \tau)$ is the cross-term ambiguity function between $d_i(t)$ and $d_k(t)$.

5. CS-BASED RECONSTRUCTION OF TIME-FREQUENCY REPRESENTATION

Consider that each signal source is a single-component AM-FM signal:

$$d_i(t) = a_i(t)e^{j\phi_i(t)}, \quad i = 1, \ldots, K,$$

where $a_i(t)$ and $\phi_i(t)$ are, respectively, the amplitude and phase of the $i$th signal.

Ideally, its time-frequency distribution should represent the total energy along time-frequency trajectories according to

$$D_{d_i d_i}(t, f) = a_i^2(t)(f - \phi_i(t)/(2\pi)).$$

In reality, however, the resolution of time-frequency distributions is highly limited by a number of factors, such as the length of available data and the applied window. This is further complicated with multiple signal arrivals. The CS-based approach seeks a perfectly localized solution that is further complicated with multiple signal arrivals. The length of available data and the applied window. This is further complicated with multiple signal arrivals. The CS-based approach seeks a perfectly localized solution that is further complicated with multiple signal arrivals. The length of available data and the applied window. This is further complicated with multiple signal arrivals. The CS-based approach seeks a perfectly localized solution that is further complicated with multiple signal arrivals. The length of available data and the applied window. This is further complicated with multiple signal arrivals. The CS-based approach seeks a perfectly localized solution that is further complicated with multiple signal arrivals. The length of available data and the applied window. This is further complicated with multiple signal arrivals. The CS-based approach seeks a perfectly localized solution that is further complicated with multiple signal arrivals. The length of available data and the applied window. This is further complicated with multiple signal arrivals. The CS-based approach seeks a perfectly localized solution that is further complicated with multiple signal arrivals. The length of available data and the applied window. This is further complicated with multiple signal arrivals. The CS-based approach seeks a perfectly localized solution that is further complicated with multiple signal arrivals. The length of available data and the applied window.

The primary constraint which is given by the above expression imposes a strict equality over $\Omega$ in the ambiguity domain. This however can be relaxed according to [16]

$$D = \arg \min_B ||B||_1 \text{ s.t. } \mathcal{F}^{-1}(B) - A = 0. \quad (12)$$

The spatial correlation matrix of the mixing matrix is given by

$$H^TH = \begin{bmatrix} 4.33 & -0.28 \pm 0.55j \\ -0.28 + 0.55j & 4.20 \end{bmatrix}.$$

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allows the compressive sensing (CS) to act on data of much higher signal-to-interference-plus-noise ratio (SINR). It was shown that this pre-processing step leads to finer resolution and more accurate representation of the nonstationary signal behavior when solving the $L_1$ norm minimization problem in the time-frequency domain.

REFERENCES


